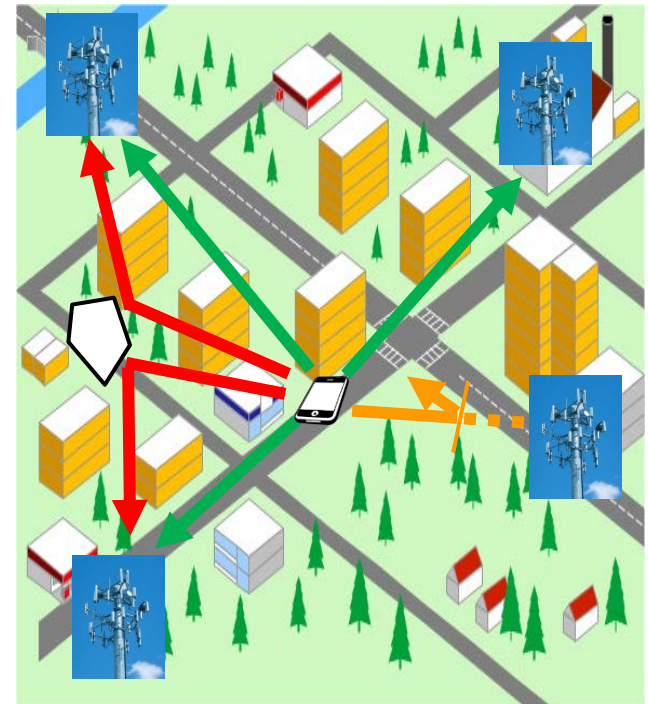


Localization of Unknown Signals over Multipath Channels

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- ▶ Goal: Localization (geolocation) of unknown RF emitters in multipath environments
- ▶ Challenges:
 - Conventional methods such as TDOA based on line-of-sight (LOS)
 - **Non-line-of-sight (NLOS) paths**
 - **Blocked LOS paths**
(e.g. indoor source)
- ▶ Applications:
 - Defense applications
 - Location based services
 - E911



Goal

- ▶ Estimate emitters locations

Sensor assumptions

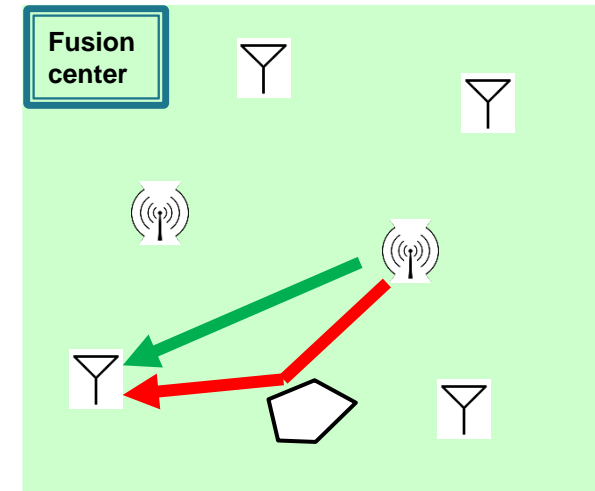
- ▶ Network of distributed sensors with fixed, known locations
- ▶ Sensors have ideal communication with a fusion center
- ▶ Sensors are time synchronized

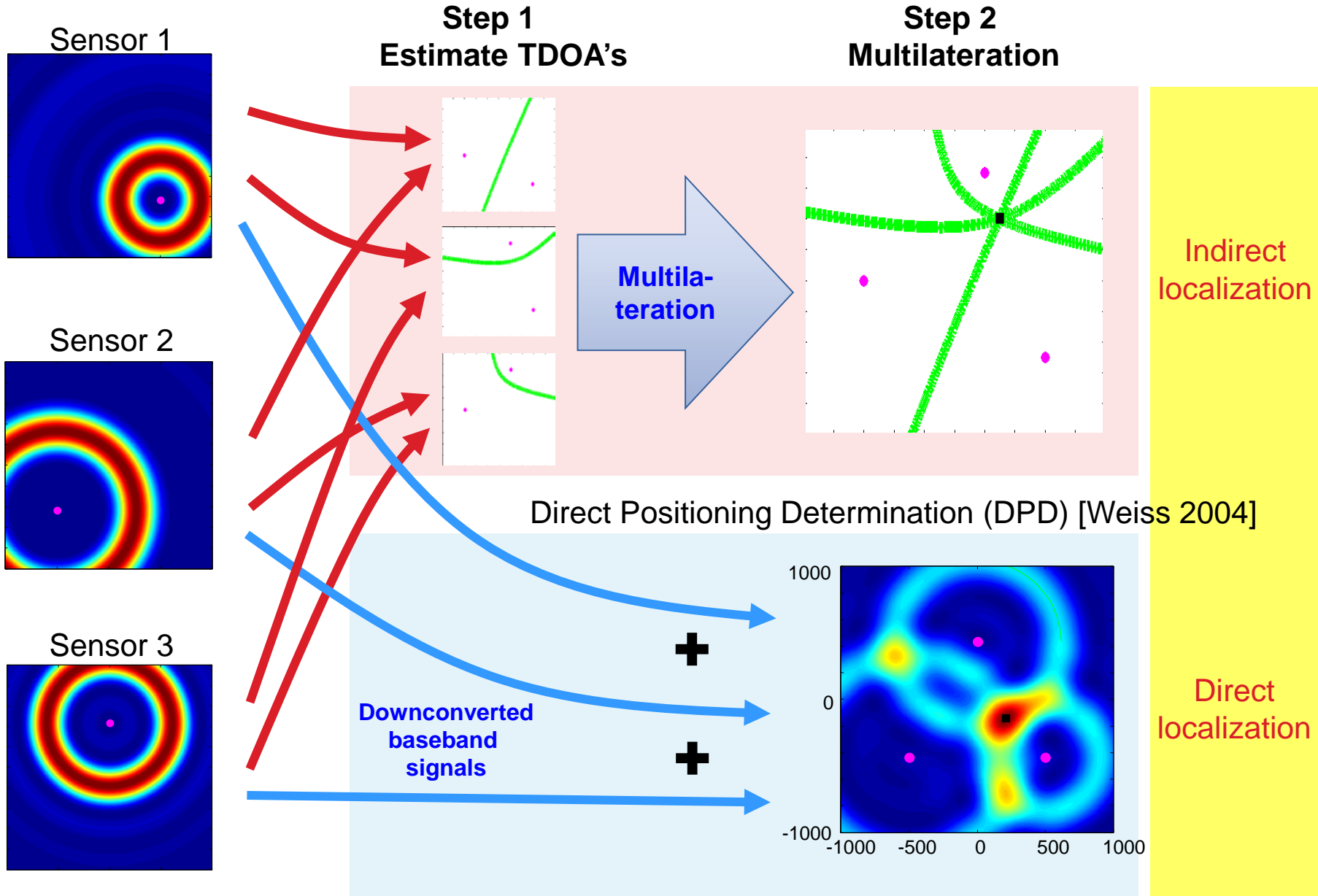
Source assumptions

- ▶ Emitter waveforms are unknown
- ▶ PSD is known

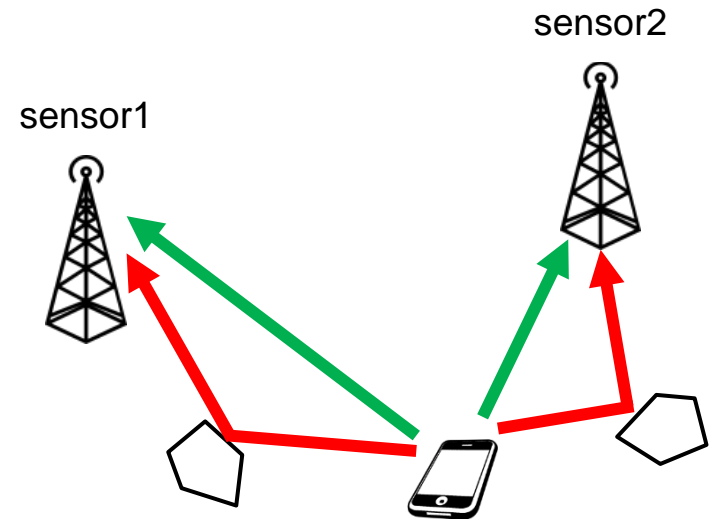
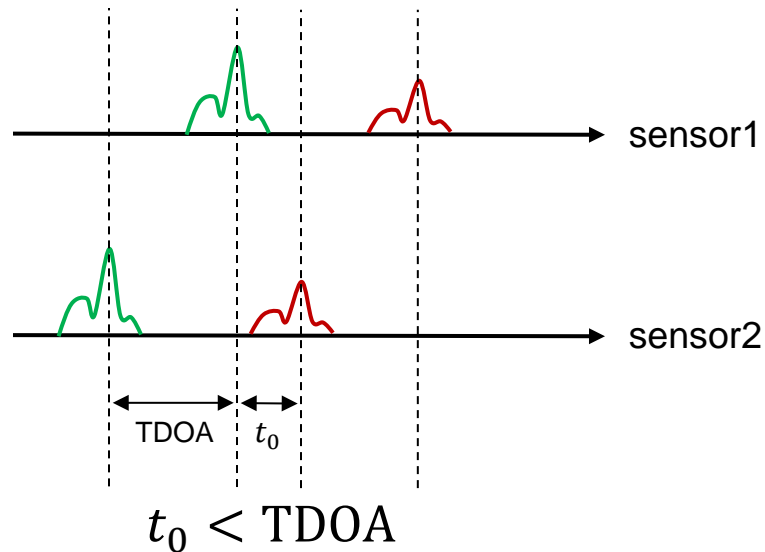
Channel assumptions

- ▶ Time-invariant unknown multipath channel
- ▶ No prior information on the multipath channel





- ▶ TDOA, DPD fail in multipath channels – can not apply the principle that the shortest delay difference = LOS TDOA
- ▶ Very scarce literature on localizing emitters over multipath channels.



- Signal at n-th sensor

$$\tilde{z}_n(t) = \underbrace{\sum_{q=1}^Q \alpha_{nq} \tilde{s}_q(t - \tau_n(\mathbf{p}_q))}_{\text{LOS}} + \underbrace{\sum_{q=1}^Q \sum_{m=1}^{M_{nq}} \beta_{nq}^{(m)} \tilde{s}_q(t - \tau_{nq}^{(m)})}_{\text{NLOS}} + \tilde{w}_n(t)$$

- N sensors, Q emitters

- \tilde{s}_q = source signal

- LOS

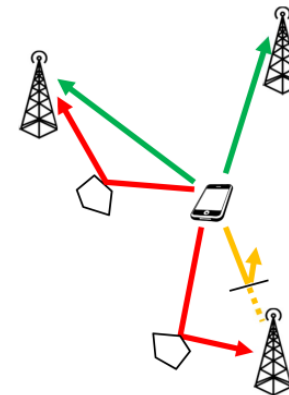
α_{nq} = complex gain LOS path

\mathbf{p}_q = source location

- NLOS

$\beta_{nq}^{(m)}$ = complex gain NLOS path

$\tau_{nq}^{(m)}$ = multipath time delay



Signal model for n-th sensor (frequency domain)

$$z_n(f) = \sum_{q=1}^Q \alpha_{nq} s_q(f) e^{-j2\pi f \tau_n(\mathbf{p}_q)} + \sum_{q=1}^Q \sum_{m=1}^{M_{nq}} \beta_{nq}^{(m)} s_q(f) e^{-j2\pi f \tau_{nq}^{(m)}} + w_n(f)$$

LOS

NLOS

Covariance matrix

$$\mathbf{R}(f) = E[\mathbf{z}(f)\mathbf{z}^H(f)]$$

Model of vectorized covariance matrix $\gamma(f) = \text{vec}(\mathbf{R}(f))$

$$\gamma_n(f) = \sum_{q=1}^Q \mathbf{x}_{nq} S_q(f) e^{-j2\pi f \tau_n(\mathbf{p}_q)} + \sum_{q=1}^Q \sum_{m=1}^{M_{nq}} \mathbf{y}_{nq}^{(m)} S_q(f) e^{-j2\pi f \tau_{nq}^{(m)}} + w_n(f)$$

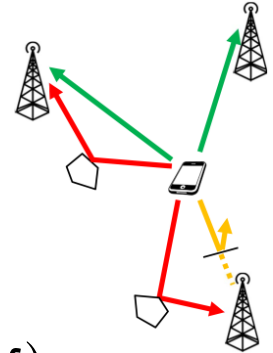
LOS

NLOS

$S_q(f)$ = known power spectral density

Covariance matrix depends on unknown parameters $\gamma(f; \boldsymbol{\theta})$

$$\boldsymbol{\theta} = \left[\mathbf{x}_{nq}, \mathbf{p}_q, M_{nq}, \mathbf{y}_{nq}^{(m)}, \tau_{nq}^{(m)} \right]$$



- MLE of unknown parameters from covariance matrix is too difficult
- Simpler alternative: approximate cov mat with sample cov mat

Compute the sample covariance matrix

$$\hat{\mathbf{R}} = \frac{1}{J} \sum_{j=1}^J \mathbf{z}_j \mathbf{z}_j^H$$

Vectorize

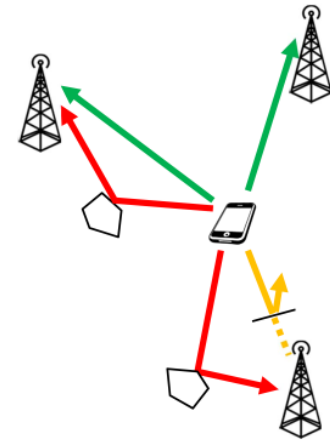
$$\hat{\boldsymbol{\gamma}} = \text{vec}(\hat{\mathbf{R}})$$

Solve optimization to find $\boldsymbol{\theta}$

$$\min_{\boldsymbol{\theta}} (\hat{\boldsymbol{\gamma}} - \boldsymbol{\gamma}(\boldsymbol{\theta}))^H \mathbf{C}^{-1} (\hat{\boldsymbol{\gamma}} - \boldsymbol{\gamma}(\boldsymbol{\theta}))$$

\mathbf{C} is the covariance matrix of the residue

$$\mathbf{C} = E \left[(\hat{\boldsymbol{\gamma}} - \boldsymbol{\gamma})(\hat{\boldsymbol{\gamma}} - \boldsymbol{\gamma})^H \right] \approx \frac{1}{J} (\hat{\mathbf{R}} \otimes \hat{\mathbf{R}})$$



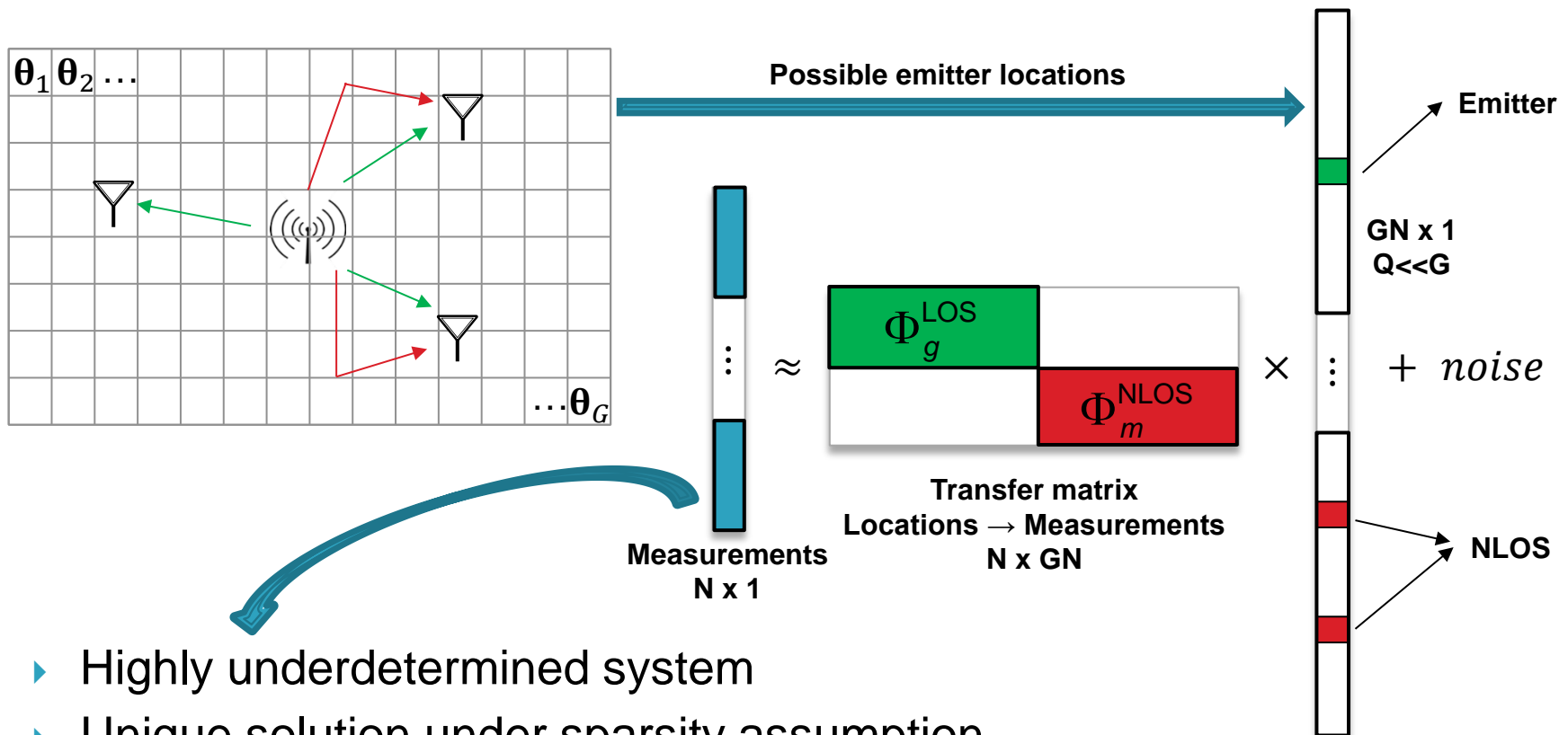
Even applying COMET leads to very complex problem

- Measurements
- Unknown parameters related to LOS paths
- Unknown parameters related to NLOS paths

$$\min_{\substack{x_{nq}, p_q \\ M_{nq} \\ y_{nq}^{(m)}, \tau_{nq}^{(m)}}} \left\| \mathbf{C}^{-1/2} \left(\hat{\gamma} - \gamma \left(x_{nq}, p_q, M_{nq}, y_{nq}^{(m)}, \tau_{nq}^{(m)} \right) \right) \right\|^2$$

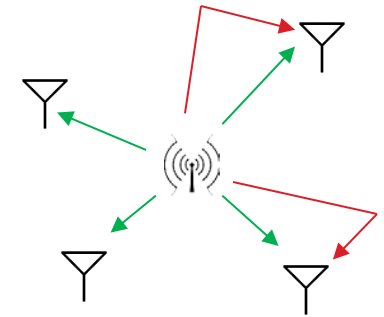
- × Large pool of unknown parameters
- × Impractical complexity

1. A small number of sources Q to be localized
2. A large number of possible locations for the sources $G \gg Q$



- ▶ Highly underdetermined system
- ▶ Unique solution under sparsity assumption
- ▶ Efficient algorithms – active area of research

- ▶ Emitters are sparse
- ▶ LOS originate from common location
- ▶ NLOS is local to the sensors
- ▶ Mixed norm optimization to control LOS vs. NLOS assignments



$$\begin{aligned} \min_{\mathbf{x}_g, \mathbf{y}_m} & \sum_{g=1}^G v \|\mathbf{x}_g\|_2 + \sum_{m=1}^M \|\mathbf{y}_m\|_1 \\ \text{subject to} & \left\| \mathbf{C}^{-1/2} (\hat{\boldsymbol{\gamma}} - \boldsymbol{\gamma}) \right\|_2^2 \leq \varepsilon \\ \boldsymbol{\gamma} &= \sum_{g=1}^G \Phi_g^{\text{LOS}} \mathbf{x}_g + \sum_{m=1}^M \Phi_m^{\text{NLOS}} \mathbf{y}_m \end{aligned}$$

Design challenge

- ▶ Explain received data with correct mixture of **LOS** and **NLOS**

$$\min_{\mathbf{x}_g, \mathbf{y}_m} \sum_{g=1}^G v \|\mathbf{x}_g\|_2 + \sum_{m=1}^M \|\mathbf{y}_m\|_1$$

- ▶ Source is missed when **LOS** is explained as **NLOS**
- ▶ False alarm occurs when **NLOS** is explained as **LOS**

Theorem

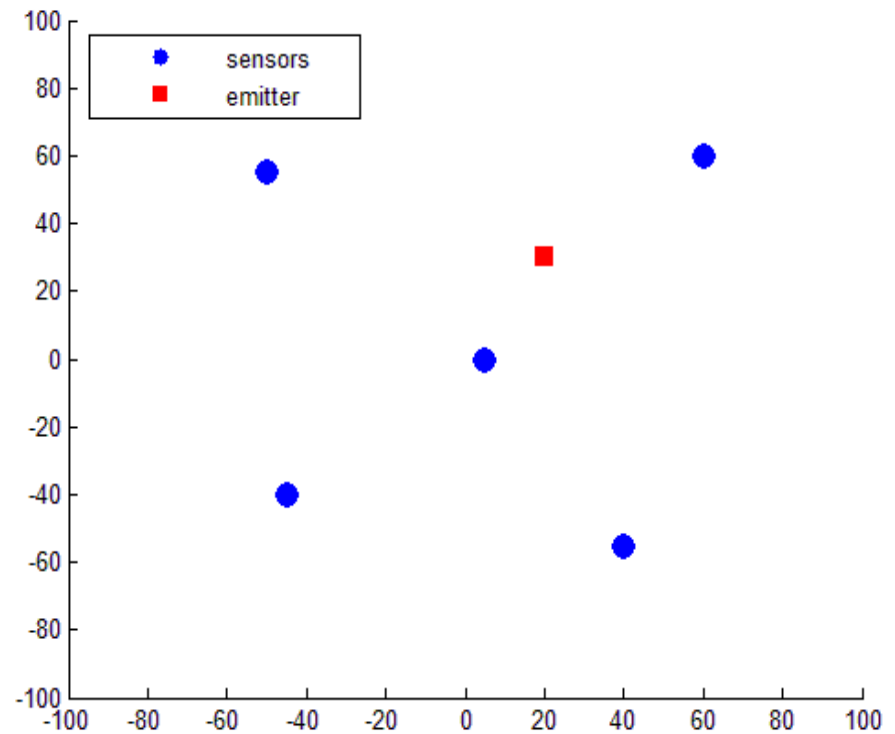
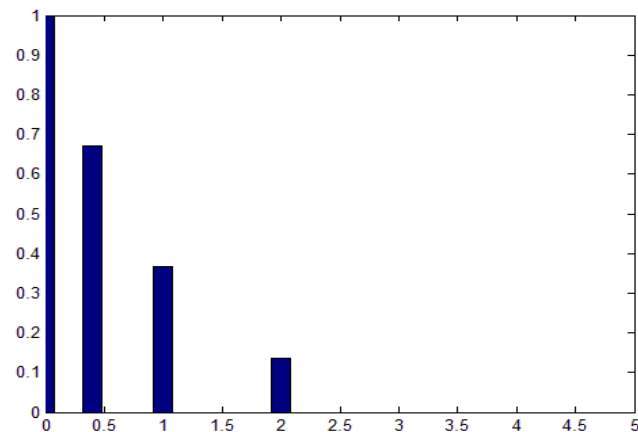
- ▶ Given measurements collected by L sensors and $L_1 < L$, if $\sqrt{L_1(L_1 - 1)} - 1 < v < \sqrt{L_1(L_1 - 1)}$, then the optimization problem will seek a feasible solution that explains a source with no less than L_1 **LOS** components

Significance

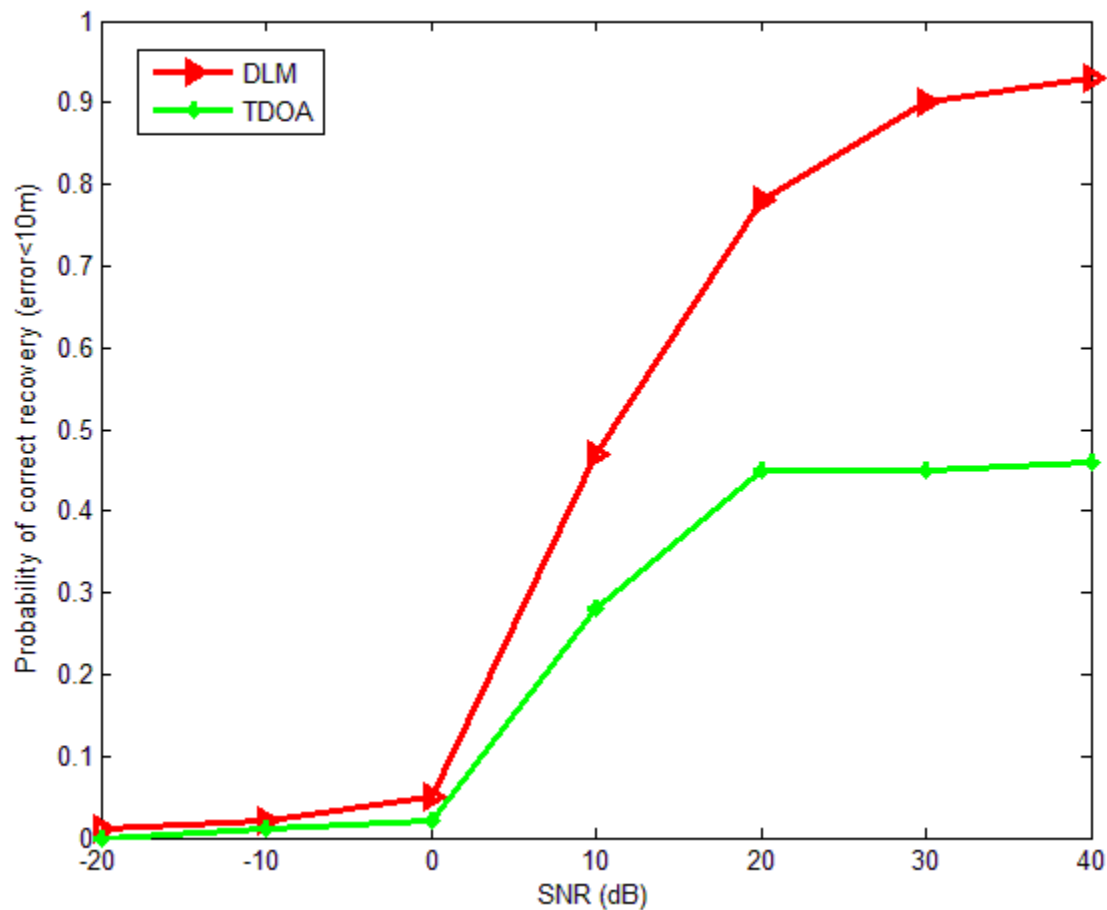
- ▶ L_1 **LOS** components are not explained as **NLOS** (prevent missed source)
- ▶ Fewer than L_1 **NLOS** components are not explained as **LOS** (prevent false alarm)

- ▶ 10 MHz emitter (30 m ranging resolution)
- ▶ Multipath channel RMS delay spread is 500 ns (exponential profile, Poisson arrivals)
- ▶ Search area: 200 x 200 m
- ▶ 5 sensor and 1 emitter
- ▶ 1000 samples/sensor

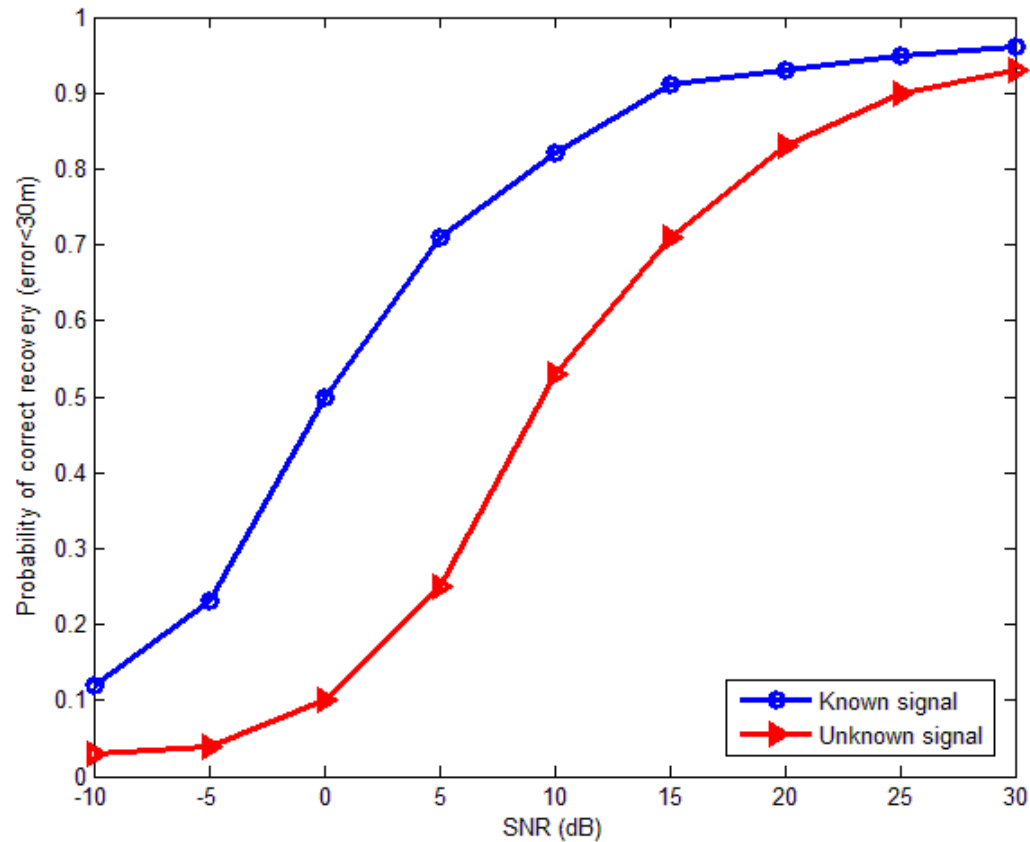
Channel impulse response



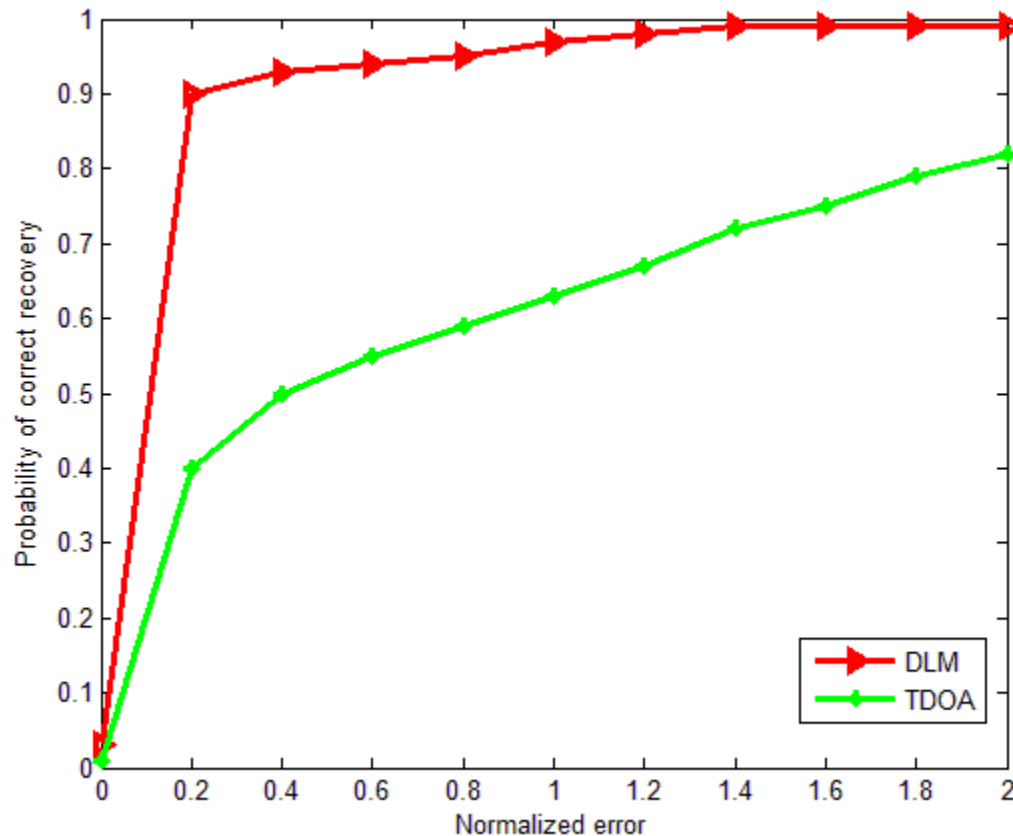
- ▶ Correct recovery if error smaller than 10 m
- ▶ Unknown signal



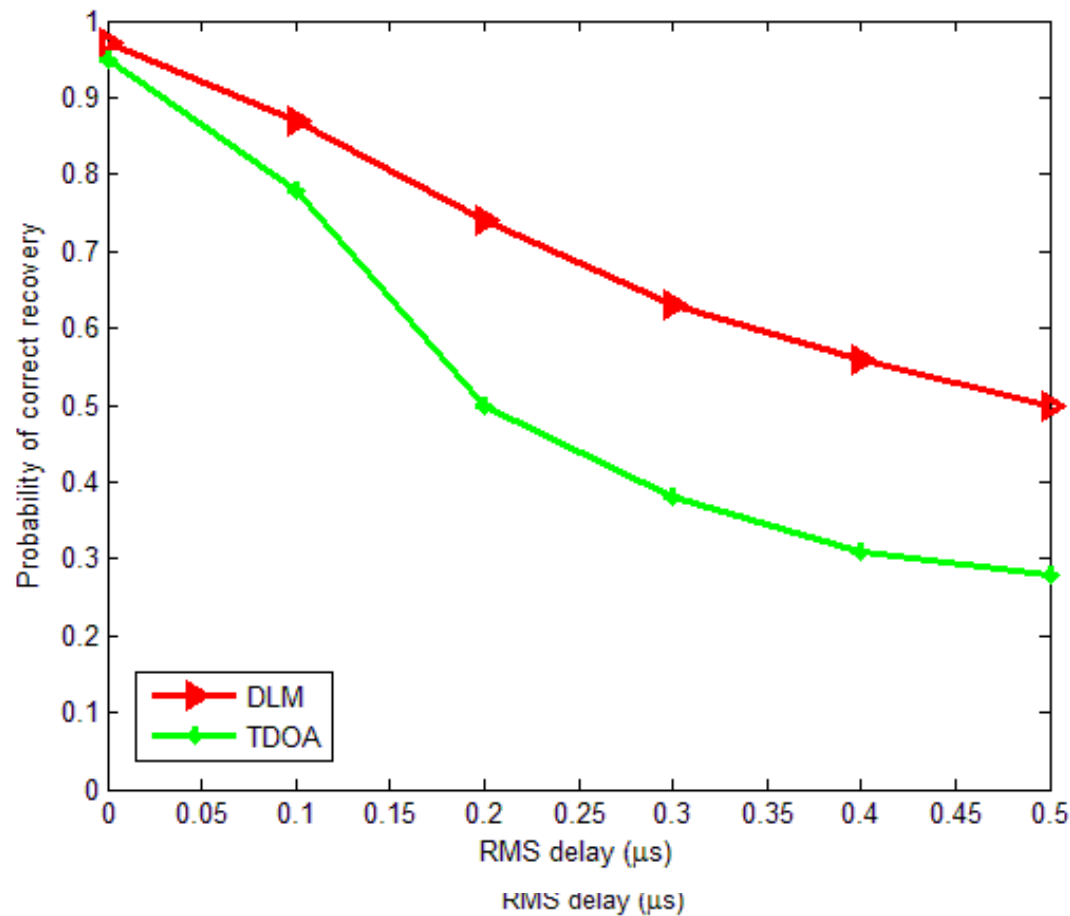
- ▶ Correct recovery if error smaller than 30 m



- ▶ Error normalized to 30m
- ▶ SNR = 30 dB per sample (1000 samples and 5 sensors)



- ▶ SNR = 30 dB per sample



- ✓ **A novel approach for localizing unknown emitters over multipath channels**
- ✓ **Solution developed directly from observations**
- ✓ **Solution relies on sparsity of emitters and of multipath**
- ✓ **Solution is blind with respect to transmitted signals and channel**
- ✓ **Mixed norm optimization exploits properties of LOS vs. NLOS**